

# Nonharmonic Fourier series

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<https://bartmcguyer.com/notes/note-10-NonharmonicFourier.pdf>

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**TL;DR:** Examples of complete orthogonal systems of sine and cosine functions that do not have a harmonic relationship.

This note presents some unusual complete sets of orthogonal basis functions that are useful in electrical problems with transmission lines and impedance matching.<sup>1</sup> These sets resemble Fourier series because they are made of sine and cosine functions. However, their wavenumbers (or eigenvalues) do not follow a harmonic progression, so they are known as nonharmonic Fourier series and sit at an interesting boundary between harmonic Fourier series and generalized Fourier series (sets of special functions). These sets can't be novel, but it seems difficult to find documentation about them.<sup>2</sup>

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## I. SETUP

Consider the ordinary differential equation

$$\frac{d^2}{dx^2} y(x) + \lambda y(x) = 0 \quad (1)$$

for the function  $y(x)$  with the domain  $x \in [x_1, x_2]$ . This could be a harmonic oscillator or the spatial portion of a 1D wave equation or the Telegrapher's equations for a transmission line (see Section V). Let its boundary conditions be of a mixed (Robin or impedance) type,<sup>3</sup>

$$c_1 y(x_1) + c_2 y'(x_1) = 0 \quad (2)$$

$$c_3 y(x_2) + c_4 y'(x_2) = 0 \quad (3)$$

where  $c_n$  are constant coefficients and a prime denotes differentiation ( $d/dx$ ).

In general, the solutions are of the form  $y(x) \propto \cos(kx + \phi)$ , with the allowed values of  $k$  and  $\phi$ —and resulting orthogonality—depending on the boundary conditions. For typical open or closed boundary conditions with each end having only one nonzero  $c_n$ , this leads to conventional Fourier series with quarter-wave or half-wave harmonic progressions for  $k$ . But, what about mixed cases in between?

Fortunately, this is a “regular” Sturm-Liouville problem,<sup>4</sup> so it is guaranteed to have a set of orthogonal eigenfunctions with positive eigenvalues  $\lambda$  that forms a complete basis over the domain. The next sections present some examples of basis sets that are controlled by a single parameter, which vary from one harmonic limit to another with anharmonic progressions in between.

Here are some overlap integrals we’ll need:

$$A_H(a, b) = \int_0^H \sin(ax) \sin(bx) dx = \begin{cases} 0 & a = b = 0 \\ \frac{H}{2} - \frac{\sin(2aH)}{4a} & a = b \neq 0 \\ \frac{a \cos(aH) \sin(bH) - b \cos(bH) \sin(aH)}{b^2 - a^2} & a \neq b, a \text{ or } b \neq 0 \end{cases} \quad (4)$$

$$B_H(a, b) = \int_0^H \cos(ax) \cos(bx) dx = \begin{cases} H & a = b = 0 \\ \frac{H}{2} + \frac{\sin(2aH)}{4a} & a = b \neq 0 \\ \frac{a \sin(aH) \cos(bH) - b \sin(bH) \cos(aH)}{a^2 - b^2} & a \neq b, a \text{ or } b \neq 0 \end{cases} \quad (5)$$

$$C_H(a, b) = \int_0^H \sin(ax) \cos(bx) dx = \begin{cases} 0 & a = b = 0 \\ \frac{\sin(aH)^2}{2a} & a = b \neq 0 \\ \frac{a - a \cos(aH) \cos(bH) - b \sin(bH) \sin(aH)}{a^2 - b^2} & a \neq b, a \text{ or } b \neq 0 \end{cases} \quad (6)$$

## II. SINE SERIES

Consider the range  $x \in [0, H]$  with the boundary conditions

$$y(0) = 0 \quad (7)$$

$$\sqrt{z/H} y(H) - \sqrt{H/z} y'(H) = 0 \quad (8)$$

where the positive, real-valued control parameter  $z \in [0, \infty]$ .

From the first boundary condition (7), the solutions must have the form

$$y(x) \propto \sin(kx) \quad (9)$$

with the allowed values of  $k$  determined by the second boundary condition.

As setup, the second boundary condition (8) becomes  $y'(H) = 0$  in the limit of  $z \rightarrow 0$ , leading to a quarter-wave Fourier sine series with the allowed values  $k_n = (2n - 1)\pi/(2H) = \pi/(2H), 3\pi/(2H), 5\pi/(2H), \dots$  for integer  $n = 1, 2, 3, \dots$ . Likewise, it becomes  $y(H) = 0$  in the limit of  $z \rightarrow \infty$ , leading to a half-wave Fourier sine series with the allowed values  $k_n = (n - 1)\pi/H = \pi/H, 2\pi/H, 3\pi/H, \dots$  for  $n = 2, 3, 4, \dots$  ( $n = 1$  is excluded to align the index for quarter- and half-wave series—see below.)

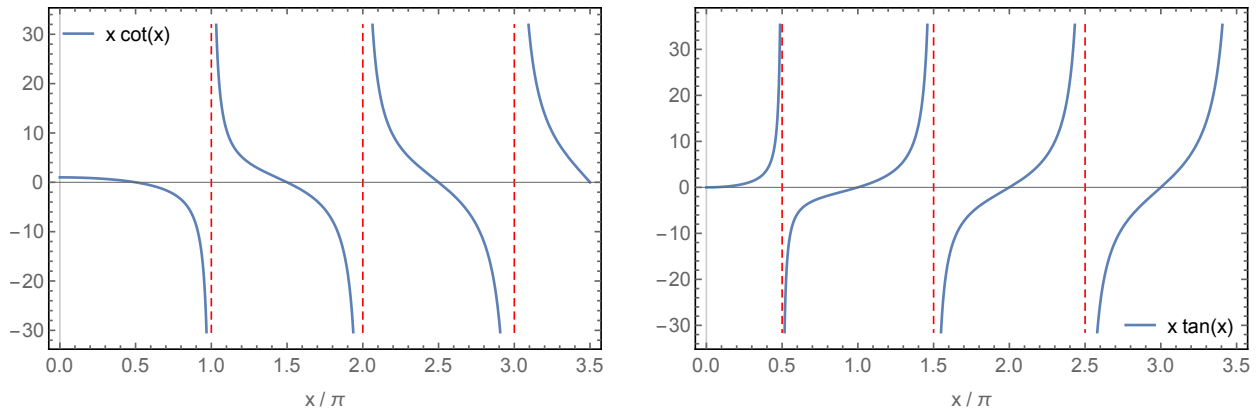


FIG. 1. Transcendental equations from the boundary conditions for the sine (left) and cosine (right) series. Note that  $\lim_{x \rightarrow 0} x \cot(x) = 1$  and  $x \tan(x) = 0$ .

For finite  $z > 0$ , the second boundary condition (8) is equivalent to

$$kH \cot(kH) = z. \quad (10)$$

The allowed values of  $k$  are given by the solutions of this transcendental equation, and correspond to the intersections between a horizontal line with value  $z$  and the curves for  $x \cot(x)$  shown in Fig. 1. Note that the limiting behaviors of  $z \rightarrow 0$  and  $\infty$  are also captured by this transcendental equation.

We can construct an orthonormal basis as the set of functions

$$s_n^{\{z\}}(x) = \frac{\sin(k_n^{\{z\}} x)}{\sqrt{A_H(k_n^{\{z\}}, k_n^{\{z\}})}} \quad (11)$$

where  $k_n^{\{z\}}$  denotes the  $n$ -th solution of the transcendental equation. (Note that the limiting case of  $k_1^{\{\infty\}} \rightarrow 0$  is excluded because  $\sin(0) = 0$ .) These basis functions are orthogonal because, for solutions of the second boundary condition,  $A_H(k_n^{\{z\}}, k_m^{\{z\}}) = 0$  for  $n \neq m$ .

Figure 2 shows the first few basis functions for different values of  $z$ , highlighting a transition from quarter-wave to half-wave harmonic limits. The numbering connects the functions between different cases, with the lowest ( $n = 1$ ) becoming absent for  $z \geq 1$ .

### III. COSINE SERIES

Consider the range  $x \in [0, H]$  with the boundary conditions

$$y'(0) = 0 \quad (12)$$

$$\sqrt{z/H} y(H) - \sqrt{H/z} y'(H) = 0 \quad (13)$$

where the positive, real-valued control parameter  $z \in [0, \infty]$ .

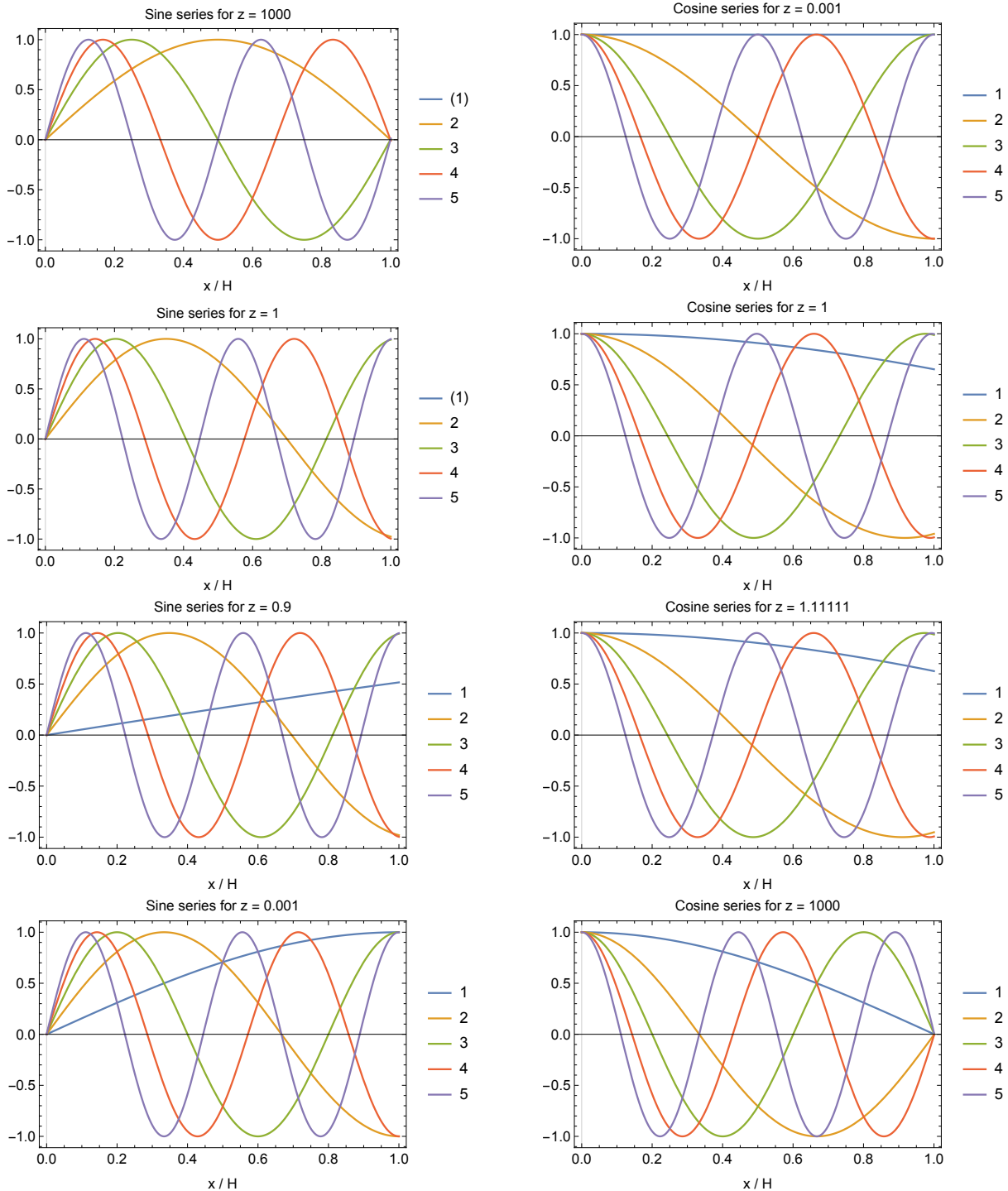


FIG. 2. First few basis functions for the sine (left) and cosine (right) series for different values of their control parameters  $z$ , transitioning from near a half-wave limit (top) to near a quarter-wave limit (bottom). The sine series loses its  $n = 1$  function as  $z$  passes through 1 from below. The basis functions are not normalized for clarity. Note that the quarter-wave limits (bottom) are mirror images of each other (left to right).

The only change from before is that the first boundary condition (12) is modified so that the solutions now have the form

$$y(x) \propto \cos(kx). \quad (14)$$

This reverses the limiting behavior for the second boundary condition (13), which now leads to a half-wave Fourier cosine series for  $z \rightarrow 0$  and a quarter-wave Fourier cosine series for  $z \rightarrow \infty$ . Unlike before, there is no abrupt removal of a function as  $z$  varies.

For finite  $z > 0$ , the second boundary condition (13) is equivalent to

$$kH \tan(kH) = z. \quad (15)$$

As before, the allowed values of  $k$  are given by the solutions of this similar transcendental equation, and correspond to the intersections between a horizontal line with value  $z$  and the curves for  $x \tan(x)$  shown in Fig. 1. Note that the limiting behaviors of  $z \rightarrow 0$  and  $\infty$  are again captured by the transcendental equation.

We can construct an orthonormal basis as the set of functions

$$s_n^{\{z\}}(x) = \frac{\cos(k_n^{\{z\}} x)}{\sqrt{B_H(k_n^{\{z\}}, k_n^{\{z\}})}} \quad (16)$$

where  $k_n^{\{z\}}$  denotes the  $n$ -th solution of the transcendental equation. (Unlike before, the limiting case of  $k_1^{\{\infty\}} \rightarrow 0$  with  $s_1^{\{\infty\}} = 1/\sqrt{H}$  is included because  $\cos(0) = 1 \neq 0$ .) Once again, these basis functions are orthogonal because, for solutions of the second boundary condition,  $B_H(k_n^{\{z\}}, k_m^{\{z\}}) = 0$  for  $n \neq m$ .

Figure 2 shows the first few basis functions for different values of  $z$ , highlighting a transition from quarter-wave to half-wave harmonic limits. The numbering connects the functions between different cases.

#### IV. DISCUSSION

These series can be readily extended to negative  $z$  with similar results. For the sine series, this gets rid of the disappearance of the null-able  $n = 1$  function about  $z = 1$ . For the cosine series, this causes the constant  $n = 1$  function present with  $z = 0$  to disappear for  $z < 0$ .

Interestingly, the sine and cosine series given above are not mutually orthogonal in general, in contrast with typical ‘‘biorthogonal’’ harmonic Fourier series. In fact,  $C_H(a, b)$  suggests that there doesn’t seem to be a way to pair anharmonic sine and cosine series to be mutually orthogonal. One way to produce a more general series would be to have mixed boundary conditions at both ends, which may produce a series with effectively both sine and cosine terms that are mutually orthogonal.

There is a similarity between these results and a ‘‘stretched’’ harmonic Fourier series that is complete over a larger domain than needed. Superficially, starting at the quarter-wave limit, varying  $z$  appears to stretch the domain of each function in these series given above

beyond  $[0, H]$ , but does so differently for each function. This stretching only aligns again harmonically in the limiting case of  $z \rightarrow \infty$ , except for  $n = 0$ . If you were to instead construct a harmonic Fourier series over a larger domain  $[0, H']$  with  $H' > H$ , the results would look roughly similar. However, the basis functions would no longer be orthogonal on the original domain  $[0, H]$ , but instead would be “over complete” on that domain. This topic is important to and discussed in Ref. 1.

## V. CONNECTION WITH TRANSMISSION LINES

We can make a connection with standing-wave resonances on transmission lines as follows. Consider a one-dimensional line with voltage  $V(x, t)$  and current  $I(x, t)$  at position  $x$  and time  $t$  along the line given by the Telegrapher’s equations

$$\frac{\partial V}{\partial x} = - \left( r + l \frac{\partial}{\partial t} \right) I \quad (17)$$

$$\frac{\partial I}{\partial x} = - \left( g + c \frac{\partial}{\partial t} \right) V, \quad (18)$$

with series resistance  $r$ , series inductance  $l$ , shunt conductance  $g$ , and shunt capacitance  $c$  per length. By taking the spatial derivative of one of the equations, we can combine the equations as either

$$\left[ \frac{\partial^2}{\partial x^2} - \left( r + l \frac{\partial}{\partial t} \right) \left( g + c \frac{\partial}{\partial t} \right) \right] V(x, t) = 0 \quad (19)$$

or another equation of the same form for  $I(x, t)$ . If we try solutions of the form  $V(x, t) = X_v(x)T_v(t)$  and  $I(x, t) = X_i(x)T_i(t)$ , then these equations separate to

$$X_v'' + \lambda_v X_v = 0 \quad (20)$$

$$X_i'' + \lambda_i X_i = 0 \quad (21)$$

$$\left[ \lambda_v^2 + \left( r + l \frac{\partial}{\partial t} \right) \left( g + c \frac{\partial}{\partial t} \right) \right] T_v = 0 \quad (22)$$

$$\left[ \lambda_i^2 + \left( r + l \frac{\partial}{\partial t} \right) \left( g + c \frac{\partial}{\partial t} \right) \right] T_i = 0, \quad (23)$$

where primes denote differentiation. The first two equations show that the spatial dependence of the current and the voltage are both of the same form as (1).

Often, the ends of a line are terminated by other circuits described by impedances. This leads to boundary conditions of the form

$$V(x_n, t) = (-1)^n Z_n I(x_n, t) \quad (24)$$

for the ends  $x_n = x_1$  and  $x_2$ , where  $Z_n$  is the terminating impedance at  $x_n$ . The sign factor accounts for the direction of the current  $I(x, t)$  being reversed at  $x_1$ .

Going back to the Telegrapher's equations, for standing-wave resonances we are interested in damped oscillatory solutions of the form  $T_v(t)$  and  $T_i(t) \propto e^{st}$  with Laplace frequency  $s = i\omega - \Gamma$ . Then (17) and (18) can be re-arrange as

$$\frac{\partial V}{\partial x} = -(r + sl)I \quad (25)$$

$$\frac{\partial I}{\partial x} = -(g + sc)V. \quad (26)$$

Either (25) or (26) can be used to replace  $I$  with  $\partial V/\partial x$  or  $V$  with  $\partial I/\partial x$  in (24), making it of the same form as the mixed boundary conditions in (2) and (3). Therefore, the basis functions for voltage or current can be nonharmonic Fourier series. Note that the impedance usually also depends on  $s$ , as in  $Z_n = Z_n(s)$ , so the orthonormal basis for voltage or current depends on the value of  $s$  in general.

For a more straightforward example of a line shorted to ground on one end and terminated by a capacitor on the other end that models an ideal Tesla coil, please see Ref. 1.

## REFERENCES

<sup>1</sup>B. H. McGuyer, "Deriving the equivalent circuit of a Tesla coil," technical note, 2020.

Available online: <https://bartmcguyer.com/notes/note-11-TcEquations.pdf>

<sup>2</sup>Please share if you know of a good reference! I haven't found them yet in standard books.

<sup>3</sup>Wikipedia, "Robin boundary condition," accessed 2020.

Online: [https://en.wikipedia.org/wiki/Robin\\_boundary\\_condition](https://en.wikipedia.org/wiki/Robin_boundary_condition)

<sup>4</sup>Wikipedia, "Sturm-Liouville theory," accessed 2020.

Online: [https://en.wikipedia.org/wiki/Sturm%E2%80%93Liouville\\_theory](https://en.wikipedia.org/wiki/Sturm%E2%80%93Liouville_theory)